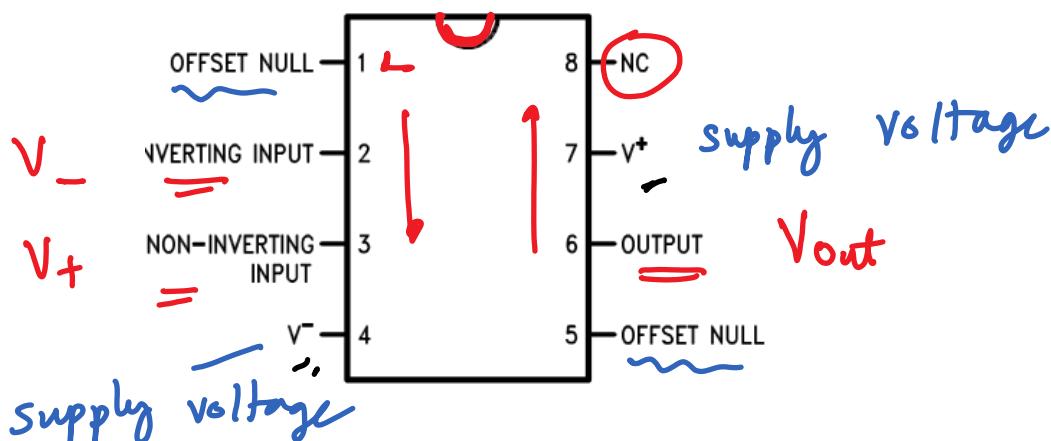


pinout

Real op-amp

Google search for LM 741 specs sheet

## ① Pin outs for LM 741

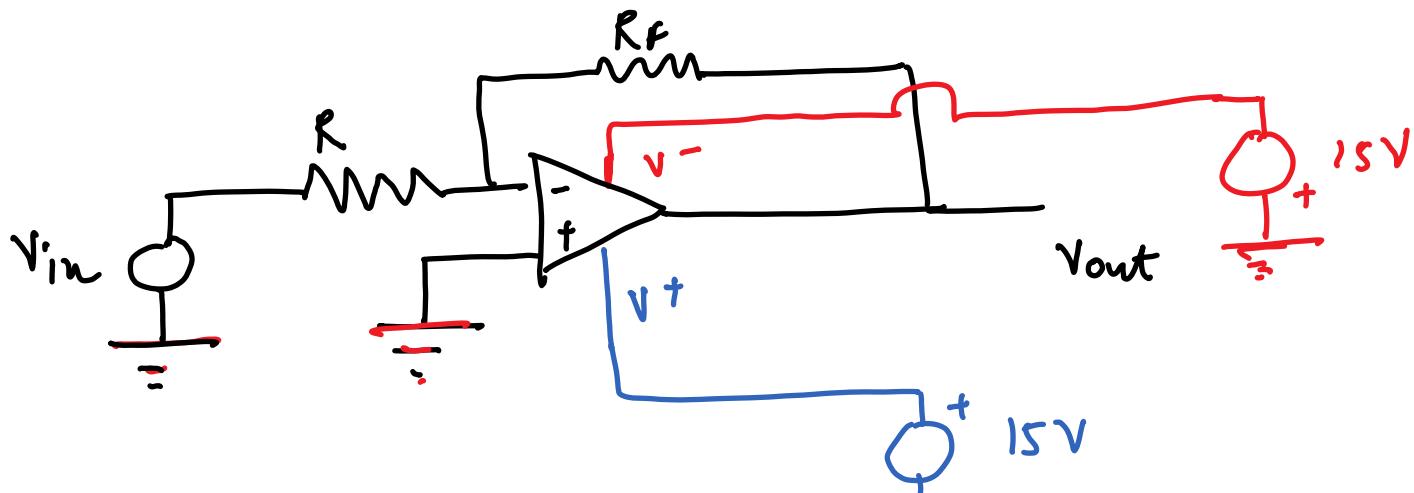


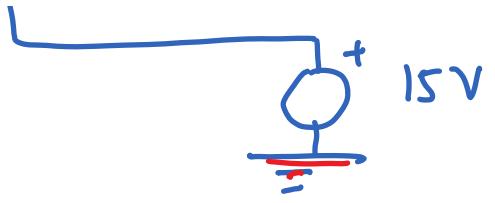
### → 6.3 Recommended Operating Conditions

over operating free-air temperature range (unless otherwise noted)

		MIN	NOM	MAX	UNIT
Supply voltage (VDD-GND)	LM741, LM741A	±10	±15	±22	V
	LM741C	±10	±15	±18	
Temperature	LM741, LM741A	-55	125		°C
	LM741C	0	70		

V<sup>+</sup> & V<sup>-</sup> to 15V & -15V respectively.





$$\frac{V_{out}}{V_{in}} = -\frac{R_f}{R}$$

We choose  $R_f, R$  such that  $R_f/R = 100$

$$\frac{V_{out}}{V_{in}} = -100$$

If  $V_{in} = 1V \Rightarrow V_{out} = -100(1) = -100V$

$\downarrow$   
theoretical

In reality

$$\begin{aligned} V_{out} &= -15 + 14 \\ &= -13.6 \end{aligned}$$

$V_{out}$  saturates to  $+13.6$  or  $-13.6$

# Max ratings

## 6.1 Absolute Maximum Ratings

over operating free-air temperature range (unless otherwise noted)<sup>(1)(2)(3)</sup>

		MIN	MAX	UNIT
Supply voltage	LM741, LM741A LM741C		±22 ±18	V
Power dissipation <sup>(4)</sup>			500	mW
Differential input voltage			±30	V
Input voltage <sup>(5)</sup>			±15	V
Output short circuit duration			Continuous	
Operating temperature	LM741, LM741A LM741C	-50 0	125 70	°C
Junction temperature	LM741, LM741A LM741C		150 100	°C
Soldering information	PDIP package (10 seconds) CDIP or TO-99 package (10 seconds)		260 300	°C
Storage temperature, T <sub>stg</sub>		-65	150	°C

$$V_{in\ (max)} = \pm 15V$$

### 6.5 Electrical Characteristics, LM741<sup>(1)</sup>

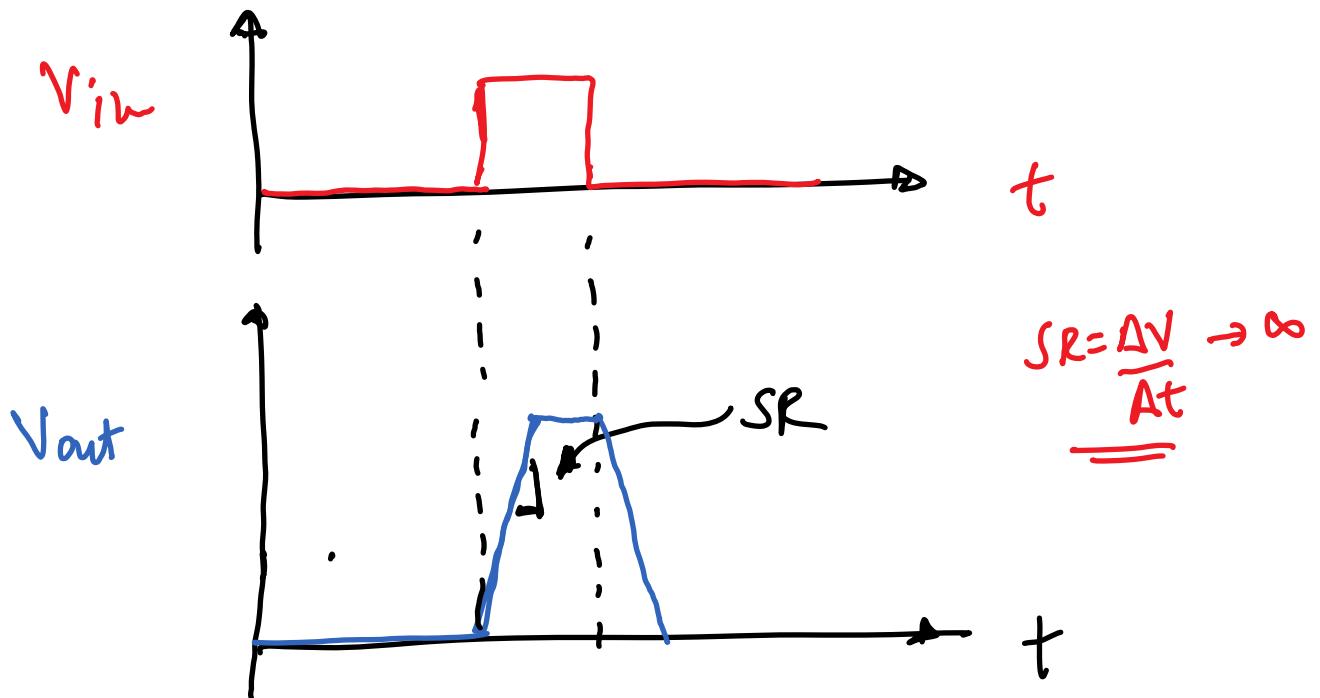
PARAMETER	TEST CONDITIONS		MIN	TYP	MAX	UNIT
Input offset voltage	$R_S \leq 10 \text{ k}\Omega$	$T_A = 25^\circ\text{C}$		1	5	mV
		$T_{A\text{MIN}} \leq T_A \leq T_{A\text{MAX}}$			6	mV
Input offset voltage adjustment range	$T_A = 25^\circ\text{C}, V_S = \pm 20 \text{ V}$			$\pm 15$		mV
		$T_A = 25^\circ\text{C}$	20	200		nA
Input offset current	$T_{A\text{MIN}} \leq T_A \leq T_{A\text{MAX}}$		85	500		
		$T_A = 25^\circ\text{C}$	80	500		nA
Input bias current	$T_{A\text{MIN}} \leq T_A \leq T_{A\text{MAX}}$			1.5		$\mu\text{A}$
		$T_A = 25^\circ\text{C}, V_S = \pm 20 \text{ V}$	0.3	2		$\text{M}\Omega$
Input resistance	$T_{A\text{MIN}} \leq T_A \leq T_{A\text{MAX}}$		$\pm 12$	$\pm 13$		V
		$V_S = \pm 15 \text{ V}, V_O = \pm 10 \text{ V}, R_L \geq 2 \text{ k}\Omega$	50	200		V/mV
Large signal voltage gain	$V_S = \pm 15 \text{ V}$	$T_A = 25^\circ\text{C}$	25			
		$T_{A\text{MIN}} \leq T_A \leq T_{A\text{MAX}}$				
Output voltage swing	$V_S = \pm 15 \text{ V}$	$R_L \geq 10 \text{ k}\Omega$	$\pm 12$	$\pm 14$		V
		$R_L \geq 2 \text{ k}\Omega$	$\pm 10$	$\pm 13$		
Output short circuit current	$T_A = 25^\circ\text{C}$		25			$\text{mA}$
		$R_S \leq 10 \Omega, V_{CM} = \pm 12 \text{ V}, T_{A\text{MIN}} \leq T_A \leq T_{A\text{MAX}}$	80	95		$\text{dB}$
Common-mode rejection ratio	$V_S = \pm 20 \text{ V}$ to $V_S = \pm 5 \text{ V}, R_S \leq 10 \Omega, T_{A\text{MIN}} \leq T_A \leq T_{A\text{MAX}}$		86	96		$\text{dB}$
		$T_A = 25^\circ\text{C}$ , unity gain	0.3			$\mu\text{s}$
Transient response	Rise time Overshoot		5%			
		$T_A = 25^\circ\text{C}$ , unity gain	0.5			$\text{V}/\mu\text{s}$
Slew rate		$T_A = 25^\circ\text{C}$ , unity gain	0.5			
Supply current	$T_A = 25^\circ\text{C}$		1.7	2.8		$\text{mA}$
		$T_A = 25^\circ\text{C}$	50	85		$\text{mW}$
Power consumption	$V_S = \pm 15 \text{ V}$	$T_A = T_{A\text{MIN}}$	60	100		
		$T_A = T_{A\text{MAX}}$	45	75		

### ③ Slope rate (SR)

$$\text{LM741} \quad SR = 0.5 \text{ V/}\mu\text{s}$$

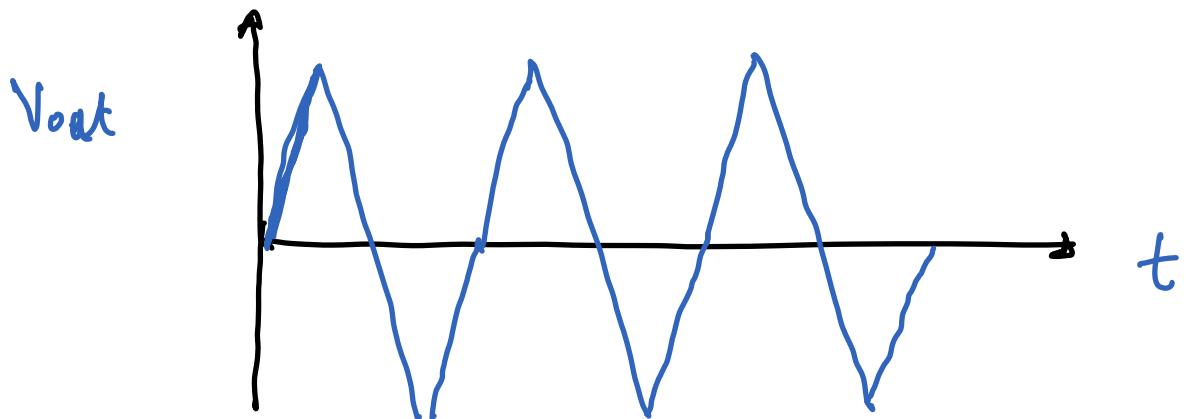
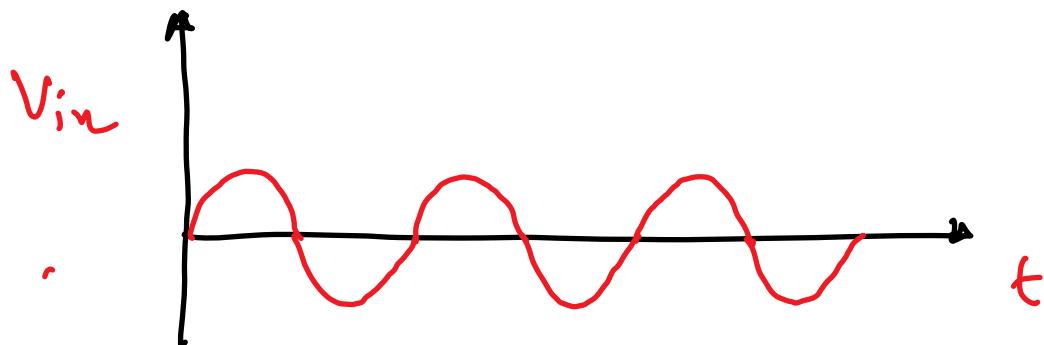
It is the rate at which the output voltage increases as a function of time

$$SR = \frac{\Delta V}{\Delta t}$$



SR is problematic with an input which is frequency dependent

e.g. sine wave



How to avoid clipping in the frequency response.

SR for a sine wave

$$SR = 2\pi f V$$

f = maximum signal frequency Hz

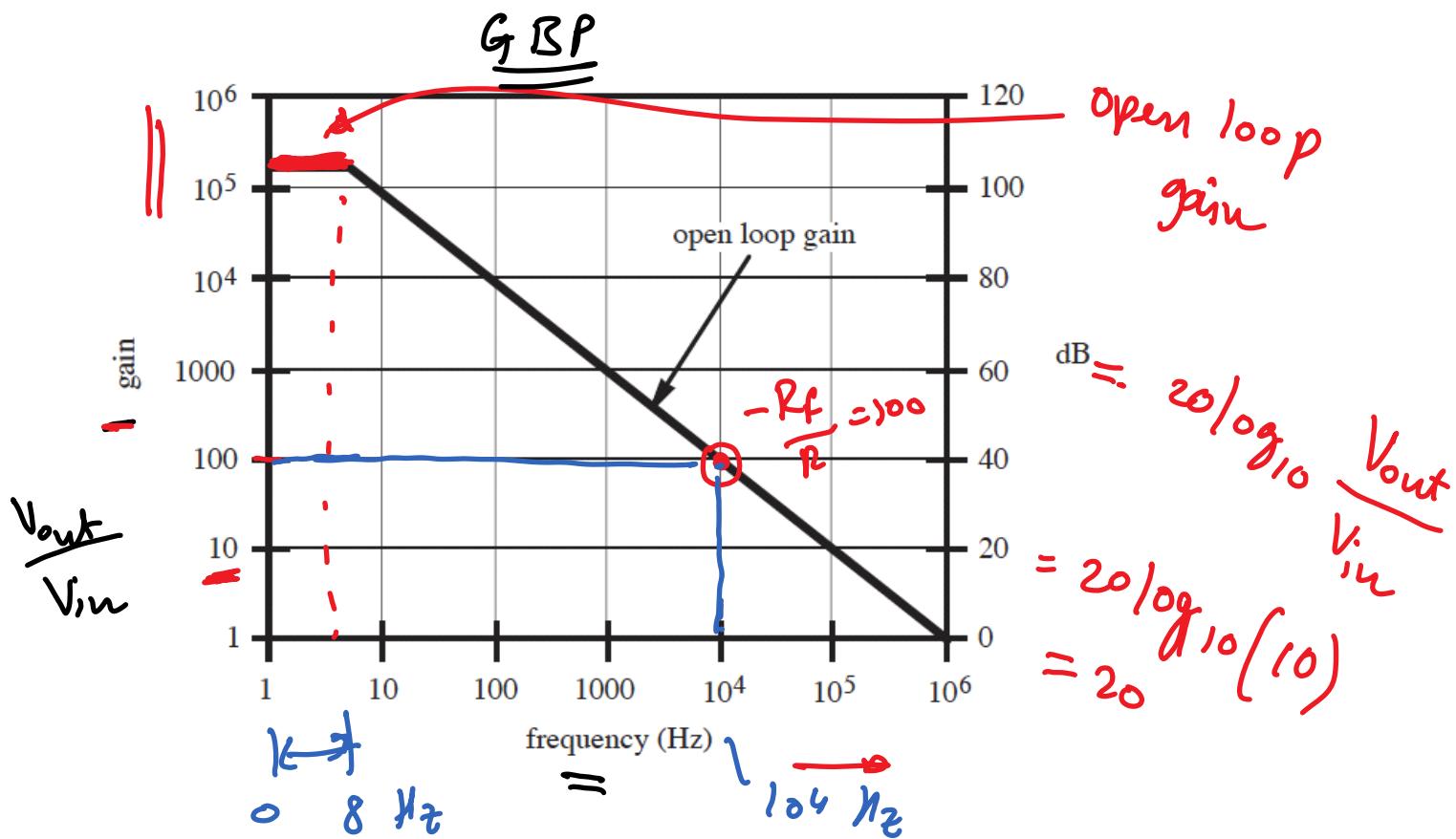
V = maximum peak voltage in V

Example: Compute the desired slew rate for an amplifier to handle a peak voltage of 10 V at a frequency of 15 kHz

$$\begin{aligned} SR &= 2\pi f V = 2\pi (15 \times 10^3) (10) \\ &= 94.2 \times 10^3 \text{ V/s} \end{aligned}$$

$$SR = 0.94 \text{ V/μs}$$

## Gain vs frequency



$$\frac{V_{\text{out}}}{V_{\text{in}}} = A_v \text{ (open loop)} \quad A_v \approx 10^5, 10^6$$

(gain)

$$\frac{V_{\text{out}}}{V_{\text{in}}} = -\frac{R_f}{R} \text{ (closed loop, inverting)}$$

### (g) Gain Bandwidth product

This is the ratio of open loop gain to the bandwidth at that gain

from specs sheet  $GSP = 200 \text{ V/V}$

$$= \frac{200}{10^3} \text{ V/V}$$

$$= \underline{\underline{2(10^5)}} \text{ V/V}$$

open loop

For inverting amplifier  $\frac{V_{out}}{V_{in}} = -\frac{R_f}{R}$

$$\underline{\underline{\approx -100}}$$

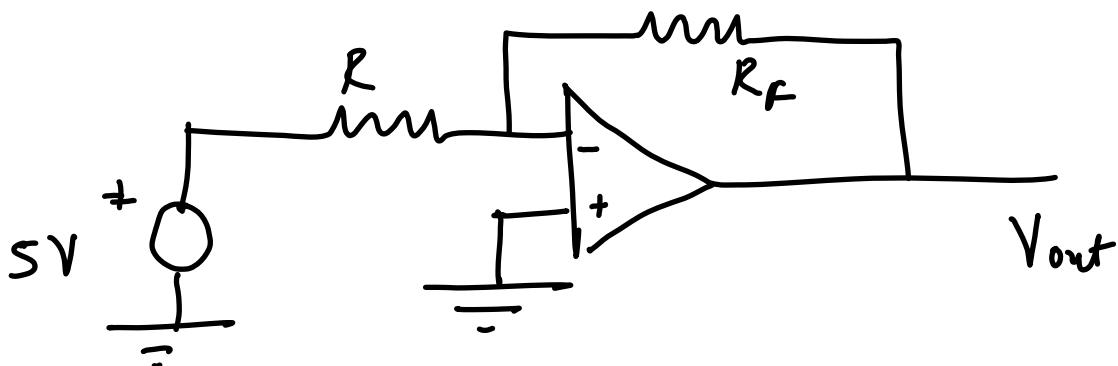
closed loop

Bandwidth for the particular  
amplifier gain is  $0 - 10^4 \text{ Hz}$   
 $0 - 10 \text{ kHz}$

⑤ Output short circuit current

$$(I_{out})_{max} \simeq 25 \text{ mA.}$$

Example on how to use this spec

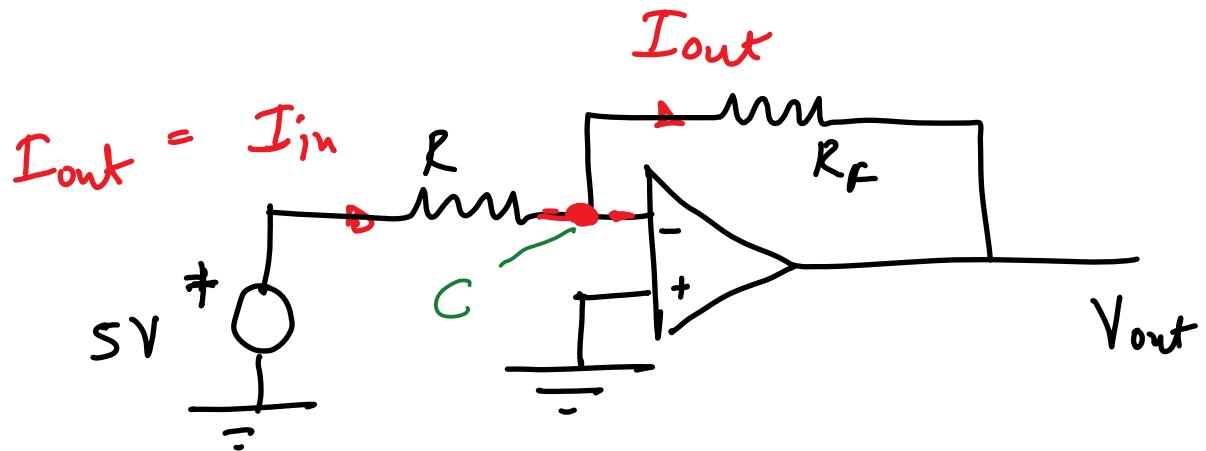


Compute the values of R<sub>f</sub>, R such that voltage gain is -2. The output short circuit current is 25 mA.

---

We derived  $\frac{V_{out}}{V_{in}} = -\frac{R_f}{R} = -2$

$$R_f = 2R$$



$$(I_{out})_{max} = 25 \text{ mA}$$

$$5 - V_C = I_{in} R = I_{out} R$$

$$V_C = V_- = V_+ = 0$$

$$R = \frac{5}{I_{out}} = \frac{5}{25 \cdot 10^{-3}} = 200 \Omega$$

$R = 200 \Omega$  (minimum)

$R_f = 2R = 400 \Omega$  (minimum)